

Integrated Photonic Resonant Modulator-Based Equalization and Optimization for DWDM

Asher Novick,^{1,*} Maarten Hattink,^{1,†} Anthony Rizzo,¹ Yuyang Wang,¹ Vignesh Gopal,¹ Songli Wang,¹ Robert Parsons,¹ and Keren Bergman¹

¹Department of Electrical Engineering, Columbia University, New York, NY, 10027, USA

[†]Now at Xscape Photonics

*asher.novick@columbia.edu

Abstract: We perform multi-wavelength signal equalization and optimization in a DWDM SiPh link by adjusting the operating regime of integrated resonant modulators. An effective increase in the optical link’s dynamic range of >3 dB is measured. © 2024 The Author(s)

1. Introduction

The next generation of optical interconnects for data center (DC) and high-performance computing (HPC) applications require unprecedented integrated optical bandwidth density and energy efficiency metrics. Recent advances in optical frequency comb sources and novel integrated photonic link architectures are meeting these metrics via dense wavelength-division multiplexing (DWDM) [1]. In such systems, the maximum total aggregated per-fiber link bandwidth is the per-channel signal bandwidth multiplied by the number of distinct carrier wavelength channels used, which is limited by the optical channel spacing and the total usable bandwidth in the optical domain. For DC and HPC applications where maintaining the optical link budget is key, spectral non-uniformity throughout the interconnect can become an obstacle towards full bandwidth utilization, requiring complex and power-hungry receiver designs with large dynamic range [2].

Integrated resonant modulators have matured as a commercially viable technology, capable of supporting broadband DWDM applications up to per channel data rates beyond 25 Gbps/λ [3]. These modulators are attractive due to their compact footprint, high modulation efficiency compatible with standard complementary metal-oxide-semiconductor (CMOS) peak-to-peak drive voltages (V_{pp}), and inherent wavelength selectivity. Fundamental to the operation of resonant modulators is the trade-off between insertion loss (IL) and signal extinction ratio (ER) for a given V_{pp} , determined by selection of the de-tuning wavelength ($\Delta\lambda$) between the optical carrier and the modulator resonance [4]. Typically, the ideal operational $\Delta\lambda$ is defined to minimize the transmission power penalty (PP_{total}) for each individual channel, disregarding the wavelength-dependent signal penalties of the interconnects.

Here, we propose utilizing the inherent relationship between a resonant modulator’s IL and ER to perform dynamic and adaptable broadband DWDM equalization and signal optimization. By operating modulators at different values of $\Delta\lambda$, wavelength channels corresponding to lower power penalties can be operated with larger modulator IL and ER, flattening the spectrum and improving average signal ER across the link and reducing crosstalk due to the relative power between victim and aggressor channels at the receiver. Through this technique, we experimentally demonstrate > 3 dB of effective increase in receiver dynamic range by dynamically tuning the resonant modulator’s $\Delta\lambda$ and V_{pp} .

2. Measurement-based Equalization Model

For on-off keying non-return to zero modulation (OOK-NRZ), the IL and ER are defined relative to the modulator optical power input (P_{in}) and power output of a “1” (P_1) and a “0” (P_0). In dB units, $IL = -10\log_{10}(P_1/P_{in})$ and $ER = 10\log_{10}(P_1/P_0)$, while the corresponding linear terms are simply $l = P_1/P_{in}$ and $r = P_1/P_0$. PP_{total} for OOK-NRZ modulation, inversely related to optical modulation amplitude (OMA), can be calculated as $PP_{total} = IL + 10\log_{10}\left(\frac{2r}{r+1}\right) + 10\log_{10}\left(\frac{r+1}{r-1}\right)$ [1]. We define a normalized OMA (OMA/P_{in}),

$$OMA/P_{in} = (P_1 - P_0)/P_{in} = l(1 - 1/r) \quad (1)$$

In the case of depletion-mode resonant modulators, IL and ER as functions of $\Delta\lambda$ and V_{pp} are derived from the DC depletion response measurements, illustrated in Fig. 1. As noted in prior works, resonant modulators suffer from self heating effects under normal operating conditions, resulting in stable performance only when $\Delta\lambda < 0$ nm. For a link comprised of a single carrier wavelength, maximizing the receiver’s photodiode peak-to-peak

photocurrent (I_{pp}) results in the highest quality signal. In the case of Fig. 1c, this corresponds to $\Delta\lambda \approx -0.05$ nm for $V_{pp} = 1.5$ V.

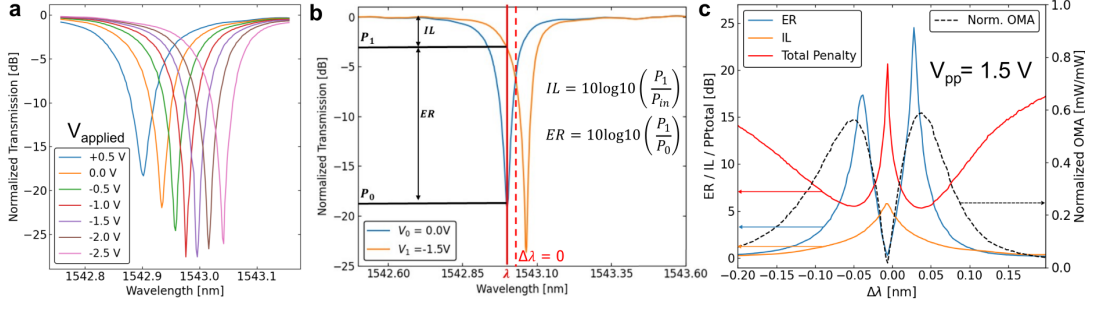


Fig. 1. **a)** Measured depletion response of an on-chip microdisk modulator. **b)** IL and ER at each wavelength is calculated based on the relative transmission of the spectra at V_0 and V_1 states. $\Delta\lambda = 0$ halfway between resonance states. **c)** ER and IL curves for $V_{pp} = 1.5$ V. PP_{total} is plotted on the same axes, while OMA/P_{in} is plotted on the right y-axis. Minimum PP_{total} corresponds to maximum OMA/P_{in} .

Instead, we consider a DWDM link consisting of many independently modulated λ s multiplexed along a single channel. Due to a combination of factors, such as a non-flat multi- λ source, wavelength-dependent optical loss (and/or gain), or even the receiver photodiode's responsivity (in A/W), spectral non-uniformity is inevitable [1]. Eq. (2) defines an operational space based on the expected I_{pp} for each carrier λ in a resonant modulator-driven DWDM system.

$$I_{pp}(\lambda, \Delta\lambda, V_{pp}) = P_{in}(\lambda) \times \frac{OMA}{P_{in}}(\Delta\lambda, V_{pp}) \times 10^{L_{link}(\lambda)/10} \times R_{PD}(\lambda) \quad (2)$$

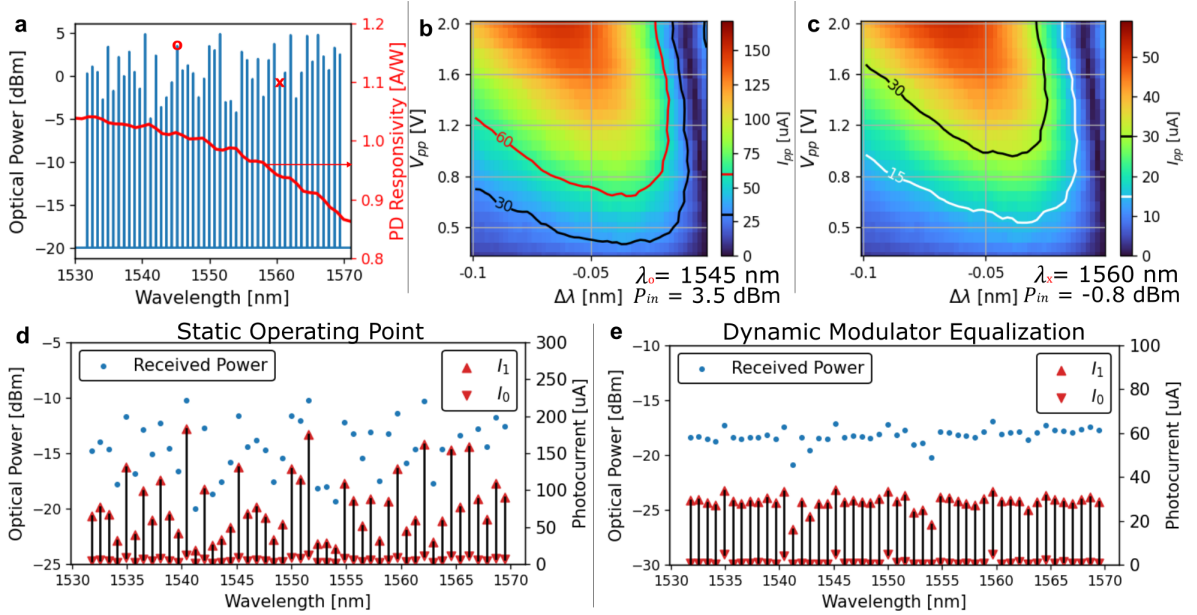


Fig. 2. **a)** Modeled DWDM source with power per λ covering 10 dB dynamic range and measured integrated photodiode responsivity at -1 V reverse bias. Assuming $IL_{link} = 10$ dB, **b)** expected $I_{pp}(\lambda_0 = 1545 \text{ nm}, \Delta\lambda, V_{pp})$ and **c)** expected $I_{pp}(\lambda_x = 1560 \text{ nm}, \Delta\lambda, V_{pp})$. Received optical power and I_{pp} operating each channel **d)** statically at the largest OMA for $V_{pp} = 1.5$ V and **e)** dynamically for $0.5 \text{ V} \in V_{pp} \in 1.5 \text{ V}$ and $-0.1 \text{ nm} \in \Delta\lambda \in 0 \text{ nm}$ to maximizing ER and I_{pp} , with minimum $I_{pp} = 15 \mu\text{A}$ and a 3 dB dynamic range limit.

Fig. 2a shows a synthesized DWDM source with 10 dB difference in power between minimum and maximum carrier wavelengths. Plotted on the secondary y-axis is a measured germanium photodiode integrated on a silicon

